



Power optimised and low latency efficient TRNG using clock gating FIFO

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ABSTRACT: An energy-efficient design strives to deliver optimum throughput while minimizing power consumption. Current research on energy-efficient processor architectures targets various abstraction levels, spanning from integration technology to algorithmic optimization. The main advantage of the proposed true random number generator utilizing programmable delay lines is to reduce correlation between several equal length oscillator rings, and thus improve the randomness qualities. Generalised FIFO is used to store generated sequence of patterns. Clock gating architecture to limit the switching activity of the address decoder which improves the power efficiency of the proposed number generator. element structure is adapted to evaluate the clock cycle to the present ring counter block and to release the clock pulse to the next ring counter block.

KEYWORDS: True random number generators, Von-Neumann correction, Linear Feedback Shift Register, Look Up table.

INTRODUCTION: Power dissipation has emerged as an important factor in the design phase of a microprocessor. Careful and intelligent design is required at different levels of computer system to obtain optimal power performance. Therefore, it is very important to know the sources of energy consumption at different levels of memory hierarchies. Since the strength of an encryption mechanism is directly related to the randomness of the binary numbers used in it, there has been an enormous need to design and develop an efficient random number generator that can produce true random numbers to implement a safe and secure cryptographic system. In addition to cyber security, random number generators (RNGs) are a vital ingredient in many other areas such as computer simulations, statistical sampling, and commercial applications like lottery games and slot machines. Random numbers are needed in some areas in computer science, such as authentication, secret key generation, game theory, and simulations. In these applications, particularly numbers should have good statistical properties and be unpredictable and non-reproducible. In modern cryptographic systems, security is based on the statistical quality and on the unpredictability of confidential keys. These keys are generated in random number generators (RNGs) using random physical phenomena that occur in the hardware devices in which the system is implemented. A widespread source of randomness in digital devices is the jitter of the clock signal generated inside the device using free running oscillators such as ring oscillators [SMS07, BLMT11, RYDV15], or self-timed rings [CFAF13]. The statistical quality and unpredictability of the generated numbers depend on the size and quality (e.g. the spectrum) of the clock jitter. It is therefore good practice to continuously monitor this jitter using an embedded jitter measurement method. As required in



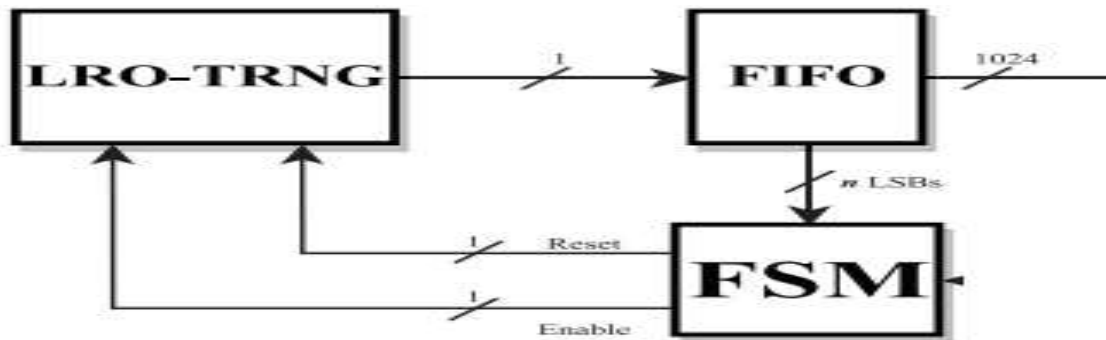
the document AIS-20/31 published by the German Federal Office for Information Security (German acronym BSI) [KS11], the measured jitter parameters should then be used as input parameters in the stochastic model used to estimate entropy, which characterizes the unpredictability of generated numbers.

LITERATURE REVIEW:

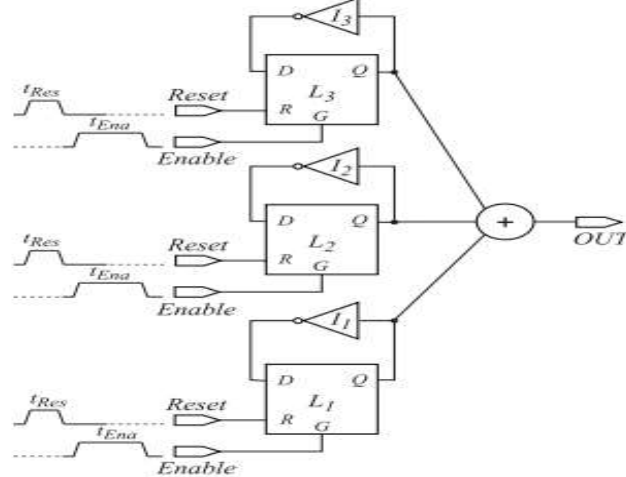
Huang et al. optimized total power consumption by efficiently designing clock gating circuit (Huang, Tu, & Li, 2012). Control logic used for clock gating has greater circuit area which is also responsible for increased dynamic power consumption. To avoid this, Integer Linear Programming (ILP) is used to optimize both the clock tree and circuit area of clock control logic. In this work, control logic circuit for clock gating is designed by reducing area requirements to 14.5% without loss of performance or increase in power consumption. Johnson et al. designed a wire-speed power processor SoC (System on Chip) based on IBM 45nm SOI (Silicon on Insulator) technology for inline processing and filtering of data with 2.3GHz core frequency, 16 cores, and 64 threads (Johnson et al., 2010). This chip uses clock gating to achieve AC power saving up to 32% (40W). Brandt et al. developed a technique for identification of speculative or passive executions which are further routed to less power consuming parts through clock gating or operand isolation for low power consumptions (Brandt, Schneider, Ahuja, & Shukla, 2010).

Koppanalil et al. designed a 1.6GHz dual-core Cortex-A9 processor using 32nm CMOS bulk process (Koppanalil, Yeung, Driscoll, Householder, & Hawkins, 2011). This chip includes power saving schemes like dynamic voltage-frequency scaling, power gating and clock gating for operation at more than GHz frequency and low power. It has all these gates introduced in its all modules to greatly reduce static and dynamic power consumption in this ARM architecture. This Cortex A9 CPU uses Clock Tree Synthesis (CTS) flow for efficient clock distribution in complete tree with hierarchical clock gating schemes.

ARCHITECTURE OF TRUE RANDOM NUMBER GENERATOR:



Block scheme of the TRNG validation testbed.



TRNG Architecture.

The block scheme of the proposed TRNG architecture is shown in Fig. 1 The single TRNG cell exploits three latches L1, L2 and L3 closed in a ring oscillator configuration through three inverters I1, I2 and I3 respectively. When a logic ‘0’ is applied to the Gate inputs (G) of the D-Latches, they are in the hold state and are insensitive to variations on the D inputs. On the other hand, when the G inputs of the latches are set to logic ‘1’, the latches become transparent and their outputs Q follow the variations on the D inputs. According to this behavior, when the latches are transparent a free running oscillation comes out, whereas, when they enter in the hold state, the logic value of the output bit is sampled. Since the propagation delay from the D inputs to the Q outputs

t_{DQi} , ($i = 1, 2, 3$) of D-Latches L1, L2 and L3, and the propagation delay t_{Ivi} , ($i = 1, 2, 3$) of the inverters I1, I2 and I3, are dependent on the physical implementation and on the delay of the routing path of the three latches, each RO exhibits an oscillation period (denoted as T_{ROi}) equals to:

$$T_{ROi} = \frac{1}{2 \cdot (t_{DQi} + t_{Ivi})} \quad (1)$$

The operation of the proposed TRNG requires the following excitation sequence: 1) Set the Reset signals to logic ‘1’ for a time t_{Res} in order to reset the outputs of the three D-Latches to logic ‘0’; 2) Set a logic ‘0’ on the Reset inputs and set the Enable signals to logic ‘1’ for a time t_{Ena} thus enabling the three ROs; 3) After a time t_{Ena} , set a logic ‘0’ on the Enable input in order to sample the output random bit. Each bit has to be generated by means of the above Reset and Enable sequence. It is evident, from these considerations, that the overall throughput (TP) is limited by the time $t_{Res} + t_{Ena}$ and the bit-sequence throughput is given by the following equation: $TP = 1 / (t_{Res} + t_{Ena})$ bits (2) As will clarified in the next sections, the entropy of the output random sequence is directly related to the accumulated jitter and therefore to the excitation time t_{Ena} of the ROs. In fact, since an increase of t_{Ena} results in an increase of the entropy and in a reduction of the throughput, the proposed TRNG architecture requires to optimize the trade-off between these two figures of merit. For this purpose, it is important to remark that the behavior of the proposed LRO-TRNG is quite different from the one of conventional RO-TRNGs. In fact, previous works exploit the jitter of the ROs and the metastability of D-flip-flops by sampling the stream started in a single instant. It has to be noticed that in the proposed LRO-TRNG, when the gates are closed (i.e., the G inputs are at logic ‘0’ and the data is sampled) a metastable state can be captured, thus increasing the entropy due to the metastability of D-Latches. However, since process, supply voltage and temperature

(PVT) variations affect the oscillation frequencies of the three LROs of a single cell in the same way (i.e., are seen as common mode variations), the XOR operation between the output of the three D-Latches L1, L2 and L3 greatly improves the resilience of the proposed TRNG to PVT variations, thus providing very good statistical performances in spite of working condition variations as will be shown in Section IV-B. Another important point to remark is that, if the oscillation frequencies of the three LROs are extremely close to each other, locking phenomena can occur [7], and the statistical performances of the TRNG can be worsened, thus requiring additional postprocessing to perform on the output sequence. To avoid these issues, the oscillation frequencies of the three LROs have to be properly unbalanced during the implementation phase by exploiting the different delays available through the different FPGA blocks and routing resources. The whole design flow and routing strategies have been investigated accordingly to [8] The LRO-TRNG has been implemented in a single FPGA Slice by means of four LUTs and three Latches. Several

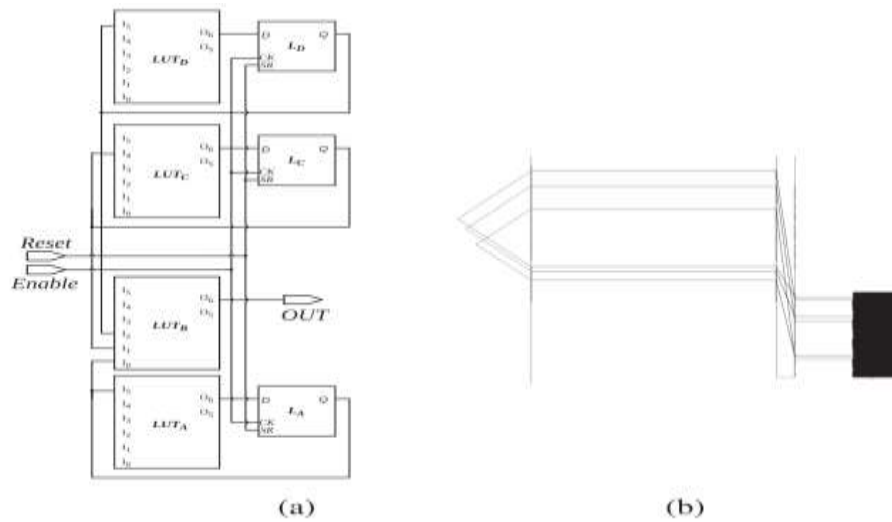
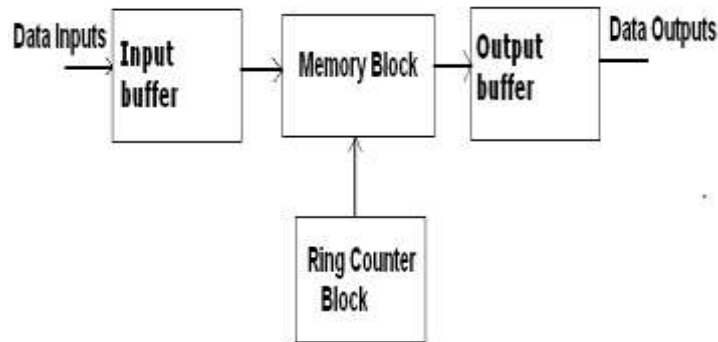
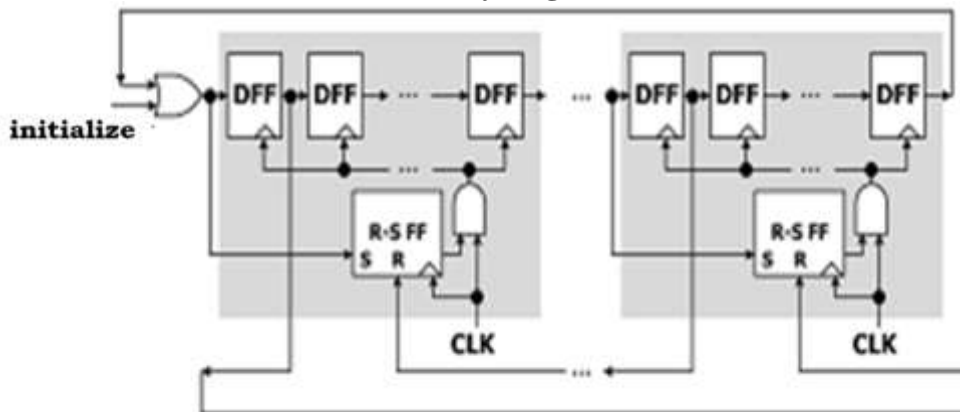


Fig.3. Slice implementation of the proposed TRNG (a); FPGA-editor's view of the Slice's intra-connections (b).routing configurations have been investigated and the one that gives the best performances in terms of randomness and entropy has been selected and depicted in Fig. 2(a). As can be observed in Fig. 2(a) the LUTA, LUTC and LUTD have been employed to implement the inverters depicted in Fig. 1, whereas the three flip-flops configured as latches are FFA, FFC and FFD. Finally the three outputs of the D-Latches are XOR-ed by means of LUTB. The FPGA-Editor floorplan view of the feedback-connections of the LRO-TRNG macro-cell is reported in Fig. 2(b). An in-depth study of the switchmatrix block and the Placement Constraints provided by the Xilinx Guide [21] allowed us to choose the intra-connections resulting in optimal randomness performance. Delays of the routed paths of LUTA, LUTC and LUTD are denoted as t_{QI5A} , t_{QI5C} and t_{QI5D} respectively. The nominal value of the delay between the LUTs' inputs and the D-Latches' outputs is about 950ps,¹ according to [2]. As can be observed the three frequencies are in the range of [300,400] MHz; in this estimation we have also taken into account the fan-in effect of the LUTB. Another important feature of the proposed implementation is the possibility for the three D-Latches to share the same Reset and Enable signals, thus improving the synchronism of the excitation sequence.

MEMORY ARCHITECTURE:



Memory Organisation



Ring Counter With SR Flip-Flops

The above block diagram shows the power controlled Ring counter. First, total block is divided into two blocks. Each block is having one SR FLIPFLOP controller to reduce constrained parameters.

RESULT:



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